



Lecture (5) Time Domain Analysis of Feedback Control Systems Performance and Characteristics -Part2

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Contents



□ Response of Second Order Systems.

<u>Second Order Systems</u>

A general second-order system (without zeros) is characterized by the following transfer function.



Second Order Systems



 $\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$

 $\Box \zeta \rightarrow$ damping ratio of the second order system, which is a <u>measure</u> of the <u>degree</u> of <u>resistance</u> to <u>change</u> in the <u>system</u> <u>output</u>.

 $\Box \omega_n \rightarrow$ un-damped natural frequency of the second order system, which is the <u>frequency</u> of <u>oscillation</u> of the <u>system without</u> <u>damping</u>

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Example 1

Determine the un-damped natural frequency and damping ratio of the following second order system.

$$\frac{C(s)}{R(s)} = \frac{4}{s^2 + 2s + 4}$$

Compare the <u>numerator</u> and <u>denominator</u> of the given transfer function with the <u>general 2nd</u> order <u>transfer</u> function.

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

 $\omega_n^2 = 4 \implies \omega_n = 2 \text{ rad/sec}$ $s^2 + 2\zeta\omega_n s + \omega_n^2 = s^2 + 2s + 4 \implies 2\zeta\omega_n s = 2s$ $\implies \zeta\omega_n = 1$

https://maxarziedu.sy/ 0.5





 \Box For the second order system described by the <u>closed</u> <u>loop</u> <u>transfer</u> function T(s), determine ω_n and ξ .

$$T(s) = \frac{C(s)}{R(s)} = \frac{24}{4s^2 + 12s + 256} = \frac{6}{s^2 + 3s + 64}$$

Compare with the standard equation we have:

Since
$$\omega_n^2 = 64$$
 and $2\zeta \omega_n = 3$ and $k\omega_n^2 = 6$

$$=> \omega_n = 8$$
 , $\zeta = \frac{3}{16} = 0.8$ and $k = 0.094$

Second Order Systems - Poles The second order system Transfer function is

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

□ The <u>characteristic</u> <u>polynomial</u> of a second order system is:

$$S^{2} + 2\zeta \omega_{n} S + \omega_{n}^{2} = (S - S_{1})(S - S_{2}) = 0$$

□ The closed-loop poles of the system are

$$s_{1}, s_{2} = \frac{-2\zeta \omega_{n} \pm \sqrt{4\zeta^{2} \omega_{n}^{2} - 4\omega_{n}^{2}}}{2} = -\zeta \omega_{n} \pm \omega_{n} \sqrt{\zeta^{2} - 1}$$

$$s_{1} = -\omega_{n}\zeta + \omega_{n} \sqrt{\zeta^{2} - 1}$$

$$s_{2} = -\omega_{n}\zeta - \omega_{n} \sqrt{\zeta^{2} - 1}$$

$$s_{3} = -\omega_{n}\zeta + \omega_{n} \sqrt{\zeta^{2} - 1}$$

$$s_{4} = -\omega_{n} \zeta \pm j\omega_{n} \sqrt{1 - \zeta^{2}}$$

$$s_{5} = -\omega_{n} \zeta + \omega_{n} \sqrt{\zeta^{2} - 1}$$

$$s_{5} = -\omega_{n} \zeta \pm j\omega_{n} \sqrt{1 - \zeta^{2}}$$

$$s_{5} = -\omega_{n} \zeta \pm j\omega_{n} \sqrt{1 - \zeta^{2}}$$

Classification of second order systems' response according to damping ratio:

- \Box <u>According</u> the value of ζ , a <u>second-order system's response</u> can be set into one of the four categories:
 - Case 1: Over damped response ($\xi > 1$) [two roots are real but not equal]
 - Case 2: Critically damped response ($\xi = 1$) [two roots are real and equal]
 - Case 3: Under damped response (0 < ξ < 1) [two roots are complex conjugate]</p>
 - Case 4: No damped (undamped) response (ξ = 0) [two roots are imaginary]

$$s_1 = -\omega_n \zeta + \omega_n \sqrt{\zeta^2 - 1}$$
$$s_2 = -\omega_n \zeta - \omega_n \sqrt{\zeta^2 - 1}$$

Classification of second order systems' response according to damping ratio:



Definition of the parameters ω_n and ζ for an underdamped, second-order system from the complex conjugate pole locations.

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$$\frac{Y(s)}{R(s)} = G(s) = \frac{K\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

The unit-step response of the transfer function is given by

$$\begin{aligned} \zeta &= 0 \\ y(t) &= K(1 - \cos \omega_n t); t \ge 0 \\ 0 &< \zeta < 1 \\ y(t) &= K \left[1 - e^{-\zeta \omega_n t} \left(\cos \omega_d t + \frac{\zeta}{\sqrt{1 - \zeta^2}} \sin \omega_d t \right) \right] = K \left[1 - \frac{e^{-\zeta \omega_n t}}{\sqrt{1 - \zeta^2}} \sin \left(\omega_d t + \tan^{-1} \frac{\sqrt{1 - \zeta^2}}{\zeta} \right) \right]; t \ge 0 \\ \zeta &= 1 \\ y(t) &= K [1 - e^{-\omega_n t} - \omega_n t e^{-\omega_n t}]; t \ge 0 \\ \zeta &> 1 \\ y(t) &= K \left[1 + \frac{1}{2\sqrt{\zeta^2 - 1}(\zeta + \sqrt{\zeta^2 - 1})} e^{-(\zeta + \sqrt{\zeta^2 - 1})\omega_n t} - \frac{1}{2\sqrt{\zeta^2 - 1}(\zeta - \sqrt{\zeta^2 - 1})} e^{-(\zeta - \sqrt{\zeta^2 - 1})\omega_n t} \right]; t \ge 0 \end{aligned}$$





Unit-Step Response of Second Order Systems Case 1: Over damped response (ξ > 1)

• The two roots of the characteristic equation s1 and s2 are real and distinct.

$$\mathbf{j}\boldsymbol{\omega} \qquad \qquad s_1 = -\omega_n \zeta + \omega_n \sqrt{\zeta^2 - 1} \\ \mathbf{v} \qquad \qquad \mathbf{v} \qquad \qquad \mathbf{v} \qquad \qquad \mathbf{v}_2 = -\omega_n \zeta - \omega_n \sqrt{\zeta^2 - 1}$$

Example: Calculate and plot the output of the system with the following transfer function:

□ Solution: With unit step input, its response is:

$$T(s) = \frac{2}{s^2 + 3s + 2}$$

$$C(s) = R(s)T(s) = \frac{2}{s(s^2 + 3s + 2)} = \frac{1}{s} + \frac{-2}{s+1} + \frac{1}{s+2}$$

□ The corresponding time domain output is given by:





Case 2: <u>Critically</u> damped response ($\xi = 1$)

• The two roots of the characteristic equation s1 and s2 are real and equal.



Example: Calculate and plot the output of the system with the following transfer function:

□ **Solution:** With unit step input, its response is:

$$T(s) = \frac{5s+4}{s^2+4s+4}$$

$$C(s) = R(s)T(s) = \frac{5s+4}{s(s^2+4s+4)} = \frac{1}{s} + \frac{-1}{s+2} + \frac{3}{(s+2)^2}$$



□ The corresponding time domain output is given by:





Case 3: Under damped response ($\xi < 1$)

• The two roots of the characteristic equation s1 and s2 are <u>complex conjugates</u> of one another.



Example: Calculate and plot the output of the system with the following transfer function:

□ **Solution:** With unit step input, its response is:

$$T(s) = \frac{4}{s^2 + 2s + 4}$$
$$C(s) = R(s)T(s) = \frac{4}{s(s^2 + 2s + 4)} = \frac{1}{s} + \frac{-s - 2}{s^2 + 2s + 4}$$

□ The corresponding time domain output is given by:





Case 4: Undamped response ($\xi = 0$)

• The two roots of the characteristic equation s1 and s2 are imaginary poles.



Example: Calculate and plot the output of the system with the following transfer function:

□ **Solution:** With unit step input, its response is:

$$T(s) = \frac{4}{s^2 + 4}$$

$$C(s) = R(s) T(s) = \frac{4}{s(s^2 + 4)} = \frac{1}{s} + \frac{-s}{s^2 + 2^2}$$

$$c(t) = L^{-1}\{C(S)\} = (1 - \cos 2t)u(t)$$

Step Response of underdamped System



Step Response of underdamped System





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Time-Domain Specification (underdamped systems)

For $0 < \zeta < 1$ and $\omega_n > 0$, the 2nd order system's response due to a unit step input looks like



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Time-Domain Specification -Delay Time

$$c(t) = 1 - \left(rac{e^{-\delta \omega_n t}}{\sqrt{1-\delta^2}}
ight) \sin(\omega_d t + heta)$$

The final value of the step response is one.

Therefore, at $t = t_d$, the value of the step response will be 0.5. Substitute, these values in the above equation.

$$egin{aligned} c(t_d) &= 0.5 = 1 - \left(rac{e^{-\delta \omega_n t_d}}{\sqrt{1 - \delta^2}}
ight) \sin(\omega_d t_d + heta) \ &\Rightarrow \left(rac{e^{-\delta \omega_n t_d}}{\sqrt{1 - \delta^2}}
ight) \sin(\omega_d t_d + heta) = 0.5 \ &t_d = rac{1 + 0.7\delta}{\omega_n} \end{aligned}$$



Time-Domain Specification – Rise Time

Rise-Time (T_r): The rise time is the time required for the response to rise from

- 10% to 90% of its final value, → over damped systems
- 5% to 95% of its final value, → Critical damped systems
- or 0% to 100% of its final value. → under damped systems





Time-Domain Specification – Rise Time

At $t = t_1 = 0$, $c(t_1) = 0$.

$$c(t) = 1 - \left(rac{e^{-\delta \omega_n t}}{\sqrt{1-\delta^2}}
ight) \sin(\omega_d t + heta)$$

$$c(t_2) = 1 = 1 - \left(rac{e^{-\delta \omega_n t_2}}{\sqrt{1-\delta^2}}
ight) \sin(\omega_d t_2 + heta)$$

$$\Rightarrow \left(rac{e^{-\delta \omega_n t_2}}{\sqrt{1-\delta^2}}
ight) \sin(\omega_d t_2 + heta) = 0$$

$$\Rightarrow \sin(\omega_d t_2 + heta) = 0$$

 $\Rightarrow \omega_d t_2 + heta = \mathrm{n}\,\pi$;n=1 For rise time

$$\Rightarrow t_2 = rac{\pi - heta}{\omega_d}$$



Time-Domain Specification – Settling time

□ The settling time (Ts): is the <u>time</u> required for the response curve to <u>reach</u> and stay within a <u>range</u> about the <u>final value</u> of size specified by absolute <u>percentage</u> of the <u>final value</u> (usually 2% or 5%).



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Time-Domain Specification – Settling time

The settling time for 5% tolerance band is: $t_s = \frac{3}{\zeta \omega_n} = 3\tau$

The settling time for 2% tolerance band is:

$$\xi_s = rac{4}{\zeta \, \omega_n} = 4 au$$

ζω,

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Time-Domain Specification – Peak Time

Peak Time (Tp): The peak time is the <u>time</u> required for the response to reach the first (<u>maximum</u>) <u>peak</u> of the <u>overshoot</u>.





Time-Domain Specification – Peak Time

$$c(t) = 1 - \left(rac{e^{-\delta \omega_n t}}{\sqrt{1-\delta^2}}
ight) \sin(\omega_d t + heta)$$

Differentiate c(t) with respect to 't'.

$$\frac{\mathrm{d}c(t)}{\mathrm{d}t} = -\left(\frac{e^{-\delta\omega_n t}}{\sqrt{1-\delta^2}}\right)\omega_d\cos(\omega_d t + \theta) - \left(\frac{-\delta\omega_n e^{-\delta\omega_n t}}{\sqrt{1-\delta^2}}\right)\sin\left(\omega_d t + \theta\right)$$

Substitute, $t=t_p$ and $rac{\mathrm{d}c(t)}{\mathrm{d}t}=0$ in the above equation.

$$\begin{split} 0 &= -\left(\frac{e^{-\delta\omega_{n}t_{p}}}{\sqrt{1-\delta^{2}}}\right) \left[\omega_{d}\cos(\omega_{d}t_{p}+\theta) - \delta\omega_{n}\sin(\omega_{d}t_{p}+\theta)\right] \\ &\Rightarrow \omega_{n}\sqrt{1-\delta^{2}}\cos(\omega_{d}t_{p}+\theta) - \delta\omega_{n}\sin(\omega_{d}t_{p}+\theta) = 0 \\ &\Rightarrow \sqrt{1-\delta^{2}}\cos(\omega_{d}t_{p}+\theta) - \delta\sin(\omega_{d}t_{p}+\theta) = 0 \\ &\Rightarrow \sin(\theta)\cos(\omega_{d}t_{p}+\theta) - \cos(\theta)\sin(\omega_{d}t_{p}+\theta) = 0 \\ &\Rightarrow \sin(\theta - \omega_{d}t_{p}-\theta) = 0 \\ &\Rightarrow \sin(-\omega_{d}t_{p}) = 0 \Rightarrow -\sin(\omega_{d}t_{p}) = 0 \Rightarrow \sin(\omega_{d}t_{p}) = 0 \\ &\Rightarrow \omega_{d}t_{p} = \pi \quad \sin(n\pi)=0; n=1 \text{ for the first peak} \\ &\Rightarrow t_{p} = \frac{\pi}{\omega_{d}} \end{split}$$



Time-Domain Specification Maximum Overshoot

Maximum Overshoot (MP): is the maximum peak value of the response curve measured from unity.

□ Maximum percent overshoot (P.O): is defined as follows:





Time-Domain Specificatio

The <u>maximum overshoot</u> is the <u>maximum peak</u> value of the <u>response curve</u> <u>measured from unity</u>. If the <u>final steady-state</u> value of the response <u>differs</u> from unity, then it is <u>common</u> to use the <u>maximum percent overshoot</u>. It is defined by

Maximum percent overshoot
$$= \frac{c(t_p) - c(\infty)}{c(\infty)} \times 100\%$$

The <u>amount</u> of the <u>maximum</u> (percent) <u>overshoot</u> <u>directly</u> <u>indicates</u> the <u>relative</u> <u>stability</u> of the <u>system</u>.



Time-Domain Specification Maximum Overshoot

$$c(t_p) = 1 - \left(rac{e^{-\delta \omega_n t_p}}{\sqrt{1-\delta^2}}
ight) \sin(\omega_d t_p + heta)$$

Substitute, $t_p = \frac{\pi}{\omega_d}$ in the right hand side of the above equation.

$$egin{aligned} c(t_P) &= 1 - \left(rac{e^{-\delta \omega_n \left(rac{\pi}{\omega_d}
ight)}}{\sqrt{1 - \delta^2}}
ight) \sin \left(\omega_d \left(rac{\pi}{\omega_d}
ight) + heta
ight) \ &\Rightarrow c(t_p) &= 1 - \left(rac{e^{-\left(rac{\delta \pi}{\sqrt{1 - \delta^2}}
ight)}}{\sqrt{1 - \delta^2}}
ight) (-\sin(heta)) \end{aligned}$$

We know that

$$\sin(heta) = \sqrt{1-\delta^2}$$

So, we will get $c(t_p)$ as

$$c(t_p) = 1 + e^{-\left(rac{\delta \pi}{\sqrt{1-\delta^2}}
ight)}$$



Time-Domain Specification Maximum Overshoot $c(t_p) = 1 + e^{-\left(\frac{\delta \pi}{\sqrt{1-\delta^2}}\right)}$

Substitute the values of $c(t_p)$ and $c(\infty)$ in the peak overshoot equation



Time-Domain Specification Maximum Overshoot

Percentage of peak overshoot % ${\cal M}_p$ can be calculated by using this formula.

$$\% M_p = rac{M_p}{c(\infty)} imes 100\%$$

By substituting the values of M_p and $c(\infty)$ in above formula, we will get the Percentage of the peak overshoot $\% M_p$ as

$$\% M_p = \left(e^{-\left(rac{\delta\pi}{\sqrt{1-\delta^2}}
ight)}
ight) imes 100\%$$

Time Domain Specifications

Rise Time

$$T_r = \frac{\pi - \theta}{\omega_d} = \frac{\pi - \theta}{\omega_n \sqrt{1 - \zeta^2}}$$

$$t_p = \frac{\pi}{\omega_d} = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}}$$

Peak Time

Delay Time

$$t_d = rac{1+0.7 \zeta}{\omega_n}$$

Settling Time (2%)

$$t_s = 4T = \frac{4}{\zeta \omega_n}$$

$$t_s = 3T = \frac{3}{\zeta \omega_n}$$

Settling Time (5%)

Maximum Overshoot

$$M_p = e^{-\frac{\pi\zeta}{\sqrt{1-\zeta^2}}} \times 100$$

Ex-1 consider a second order system with closed loop transfer function G(s) = 25 Determine time domain specifications and output response y(t) for step input. $\frac{501^{n}}{25}$ We have, $G(s) = \frac{25}{s^2 + 8s + 25}$ comparing denominator of 6(s) with s2+23wnstwn2 we get, Wn=25 > wn= 5rad/sec 25,00 = 8 = 0.8 Wd = Wn VI-82 = 5 VI- (0.8) = 3 rad/sec td = 1+0.79 = 1+0.7(0.8) = 0.312 sec ty = 1-0 = 1- cos = 1-0.6435 = 0.8327 sec tp = # = # = 1.0472 sec ts = $\frac{4}{8W} = \frac{4}{4} = 1 \text{ sec} (27, \text{ band})$ = 3 = 3 = 0.75 see (57. band) Mp = e-TTS/VI-4 = 0.0152 = 1.527.



$$Y(t) = 1 - \frac{e^{-\frac{4}{3}}}{\sqrt{1-\frac{4}{2}}} \sin(\frac{4}{3}t+e)$$

= $1 - \frac{e^{-4t}}{0.6} \sin(5t+0.6435^{c})$
= $1 - 1.6667e^{-4t} \sin(3t+0.6435^{c})$



For the control system shown in Figure, determine k and a that satisfies the following requirements:

- a) Maximum percentage overshoot P.O =10%.
- b) The 5% settling time $t_s = 1$ sec.

□ **Solution:** The closed loop transfer function is given by

$$\frac{C(s)}{R(s)} = \frac{\left(\frac{k}{s+a}\right)\left(\frac{1}{s+2}\right)}{1+\left(\frac{k}{s+a}\right)\left(\frac{1}{s+2}\right)} = \frac{k}{(s+a)(s+2)+k} = \frac{k}{s^2+s(a+2)+(2a+k)}$$



□ The maximum percent overshoot (P.O) is given by:

$$P \cdot O = e^{-\frac{\zeta \pi}{\sqrt{1-\zeta^2}}} = \frac{1 \ 0}{1 \ 0 \ 0} \rightarrow \zeta = 0.6$$

□ For 5%, the settling time ts is given by:

$$t_s = 3\tau = \frac{3}{\zeta \omega_n} = 1$$
 \longrightarrow $\omega_n = 5$

$$a+2=2\zeta\omega_n$$
 & $2a+k=\omega_n^2$

□ From these two equations we get a + 2 = 6 then a = 4 and k = 17

Ex-2 Consider unity feedback system with open lop transfer function G(s) = 16 . Determine time domain specifications of closed loop system. $\frac{501^{n}}{100}$ We have $6(s) = \frac{16}{5(s+24)}$, H(s) = 1 $G_{cl}(s) = \frac{16}{s^2 + 4s + 16}$ wh = 16 = wh = 4 stad/see 2500 = 4 > 3=0.5 Wd = wor 1-82 = 4 VI-(0.52= 3,4641 rad/sec td = 1+ 0.79 = 1+0.7 (0.5) = 0.3375 sec tr = T- 0 = T- (05-10.5 = 0.6046sec tp = TT = TT = 0.9069 Sec $t_{s} = \frac{4}{4w_{p}} = \frac{4}{2} = 2sec$ Mp = e -TT &/ VI-42 = 0.16 30 = 16.30%





Ex-3 Consider unity feed back system with open loop transfer function G(s) = 12 Determine all time domain specifications and step response of closed loop system. $\frac{501^{n}}{(s+1)(s+4)}$ we have $G(s) = \frac{12}{(s+1)(s+4)}$, H(s) = 1 $G_{cL}(s) = \frac{G(s)}{1+G(s)H(s)} = \frac{12}{s+5s+16}$ wo= 16 > wo = 4 rad/see 2800 = 5 \$ \$ = 0.625 Wd = Wn VI- 52 = 4 VI- (0.625)2 = 3. 1225 rad/see td = 1+0.7% = 1+0.7(0.625) = 0.3594 See $t_r = \frac{TT - \Theta}{W_1} = \frac{TT - CHS^{-1}S}{W_1} = 0.7193 \text{ sec}$ tp = II = 1.0061 sec ts = 4 = 4 = 1.6 sec (27. band) Mp = e-TTS/VI-42 = 0.0808 = 8.08%



k=0.75



 $Y(t) = \frac{3}{4} \left[1 - \frac{e^{-8}\omega_0 t}{\sqrt{1-8^2}} \sin(\omega_0 t t 0) \right]$ $=\frac{2}{4}\left[1-\frac{e^{-2.5t}\sin(8.1225t+0.8957^{2})}{0.7806}\right]$

	Ex4 A closed loop system is represented by the differential equation y+6y+2sy=2s T. Determine its time domain
Example 5	sol ^C we have $\dot{y} + 6\dot{y} + 25y = 25\tau$ System input is r
	Taking L.T. on both sides we get
	$s^2 \gamma(s) + 6 s \gamma(s) + 2 s \gamma(s) = 25 R c)$
	$\frac{Y(s)}{R(s)} = \frac{25}{s^2 + 6s + 25}$
	wn=25 > wn=5 rod/see
	$25\omega_n = 6 \Rightarrow s = 0.6$
	wd = wo VI-52 = 5 VI-(0.62 = 4 rad)sec
	$t_d = \frac{1+0.7 s_1}{\omega_n} = \frac{1+(0.7)(0.6)}{5} = 0.284 see$
	$t_{r} = \frac{T - \theta}{w_{d}} = \frac{T - \cos^{-1} 0.6}{4} = 0.5536 \text{sec}$
	tp = TT = # = 0.7854 sec
	ts = $\frac{4}{5w_{0}} = \frac{4}{3} = 1.3333$ sec (25. band)
	$M_P = e^{-11} \sqrt{1-4^2} = 0.0948 = 9.487$

Ex. - S consider a system with closed loop transfer function G(s) = ke . Determine k, and ke such that peak overshoot is 15% and peak time is 2 seconds. Also determine delay time, rise time and settling time. soit we have Mp=15%, top= 2see Mp= e-118/11-92 0.15 = e-119/ 11-92 -TTS = - 1.8971 VI-62 = 0.6039 - q2 1-82 = 0.3647 42 = 0.3647 - 0.3647 42 92 = 0.3647 = 0.2672 8 = 0.5170 tp = # = 2

$$w_{d} = \prod_{i=1}^{n} = 1.5708$$

$$w_{0} = \prod_{i=1}^{1.5708} = 1.8350 \text{ mod/sec}$$

$$w_{0} = \frac{1.5708}{\sqrt{1-0.2672}} = 1.8350 \text{ mod/sec}$$
From given T.F.

$$k_{1} = w_{0}^{2} = 3.3671$$

$$k_{1} = 25w_{0} = 2(0.5130)(1.8350)$$

$$= 1.8974$$

$$t_{d} = \frac{1+0.79}{w_{0}} = 0.74421 \text{ sec}$$

$$t_{T} = \frac{\Pi-\Theta}{w_{0}} = \frac{\Pi-\cos 0.5170}{1.5708} = 1.3459$$

$$sec$$

$$t_{5} = \frac{4}{5w_{0}} = \frac{4}{(0.5170)(1.8350)} = 4.2163$$